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IN THE UNITED STATES PATENT AND TRADEMARK OFFICE

In re Patent Application of:  
RAYNOR ET AL.

Serial No. 09/993,387

Filing Date: November 16, 2001

For: SOLID STATE IMAGING DEVICE  
AND ASSOCIATED METHODS

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Sir:

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Respectfully submitted,

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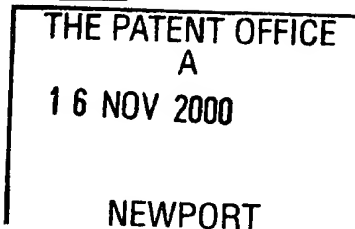
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0027931.5

16NOV00 E584113-9 D02884  
P41/7700 0.00-0027931.5

3. Full name, address and postcode of the or of  
each applicant (underline all surnames)

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Patents ADP number (if you know it)

If the applicant is a corporate body, give the  
country/state of its incorporation

United Kingdom

8022873001

4. Title of the invention

"Solid State Imaging Device"

5. Name of your agent (if you have one)

Murgitroyd & Company

"Address for service" in the United Kingdom  
to which all correspondence should be sent  
(including the postcode)

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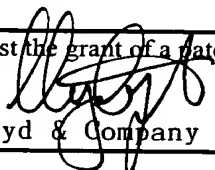
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1     **"Solid State Imaging Device"**

2

3     This invention relates to a solid state imaging  
4     device which can be operated to provide an improved  
5     shutter function.

6

7     There are various basic CMOS pixel structures. One  
8     common type, with 3 transistors per pixel, is  
9     described in US 4,407,010 ("CMOS 3T" pixel),  
10    illustrated in Fig. 1 of the accompanying drawings.  
11    This is an efficient structure as a transistor M1  
12    amplifies the photodiode output inside the pixel.  
13    Transistor M2 serves to reset the voltage on the  
14    pixel. Transistor M3 is a multiplex transistor - it  
15    enables many pixels in a column to be wired together  
16    and only one pixel enabled at a time. The device  
17    "Iload" is typically a sense amplifier which both  
18    provides a load for the source follower transistor M1  
19    and also measures the output voltage.

20

1 The typical voltage on a photodiode is shown in  
2 Figure 2. At point "1", the pixel is reset by  
3 turning on transistor M2 which sets the voltage on  
4 the reverse-biased diode to a preset voltage (VRT).  
5 After this point, light falling onto the pixel will  
6 create photo-generated electrons which will be  
7 attracted to the photodiode. This will cause the  
8 diode to be discharged. The amount of discharge is  
9 proportional to both the amount of light and also the  
10 amount of time. After a period of time (integration  
11 period, "Tint") the voltage on the pixel is measured.  
12 If the time "Tint" is kept constant, the swing will  
13 be proportional solely to the amount of light falling  
14 on the pixel.

15  
16 Typically, as shown in Figure 3, the pixels are  
17 arranged into a 2-dimensional grid of rows and  
18 columns. There is one "Iload"/sense amplifier per  
19 column. The amplifier measures the output voltage of  
20 the pixel. Several (usually all pixels in a column)  
21 share a single sense amplifier. Because of this  
22 structure all the elements in a row are read out  
23 simultaneously (into the sense amplifiers) and the  
24 rows are addressed sequentially.

25  
26 As the rows are read out sequentially, they must also  
27 be reset sequentially. This keeps the integration  
28 time "Tint", constant for the whole sensor, and the  
29 brightness of the image constant over the image  
30 plane.

31

1 This operation is called "rolling blade shutter" and  
2 is analogous to how a physical shutter in a 35mm SLR  
3 camera works. In the CMOS 3T sensor, the integration  
4 time is variable - this is achieved by varying the  
5 time between the reset and readout pulse. This is  
6 also similar to how 35mm SLR cameras work - the  
7 shutter blades move over the film at a constant rate,  
8 but a gap between the blades is adjusted to adjust  
9 the effective shutter speed.

10

11 Another common type of CMOS pixel has 4 transistors.  
12 There are various types of implementation, one of  
13 which is shown in Figure 4. The advantage of this  
14 design is that it has two storage capacitances per  
15 pixel. Cpd is formed by the "parasitic" capacitance  
16 of the photodiode. The storage node, Csn is formed  
17 partly by the stray capacitance of M1, M2 but also by  
18 creating a storage device inside the pixel. One  
19 advantage of a 4T pixel is sensitivity:  $V=Q/C$ ; hence,  
20 by reducing the value of Csn, the output voltage for  
21 a given photocharge is increased.

22

23 The 4T pixel has another advantage - its ability to  
24 form an "electronic shutter". Although arrays of  
25 either 3T or 4T pixels can be reset simultaneously,  
26 the sequential readout mechanism of the 3T pixel  
27 prevents simultaneous readout. The 4T pixel does not  
28 suffer from this problem as it has a storage element  
29 incorporated inside each pixel ("Csn" in Figure 4).  
30 This permits the entire array to be "sensed"  
31 simultaneously, i.e. photo-generated charge is

1 transferred from each pixel's Cpd to the pixel's Csn  
2 simultaneously. The readout mechanism then proceeds  
3 in a row sequential fashion, similar to the mechanism  
4 used in 3T pixels. As all the pixels in the array  
5 are reset and measured simultaneously, the array  
6 captures a "snapshot" of the light pattern falling on  
7 the sensor (unlike the "rolling blade shutter" of 3T  
8 pixels). This technique is of great value for hand-  
9 held operation of the camera as the effect of camera  
10 shake is reduced as the total time for which the  
11 array is collecting light (as opposed to the time for  
12 which an individual pixel is collecting light) is  
13 minimised.

14

15 There are significant disadvantages with a 4T pixel:

16

- 17 • the extra circuitry (M4, Csn) occupies area on the  
18 pixel and this reduces the amount of light  
19 reaching the photodiode.
- 20 • transferring all the charge from Cpd to Csn is  
21 difficult to achieve. Special CMOS manufacturing  
22 techniques are often employed to change the  
23 structure of the photodiode Cpd or the transfer  
24 transistor M4. These manufacturing techniques are  
25 very costly (as they are non-standard) and are  
26 also difficult to achieve reliably.

27

28 There are also some "linear arrays" (see Figure 5)  
29 with two rows of pixels which have separate  
30 electronics on both top and bottom. These structures  
31 are limited to a maximum of two rows.



1

2 Other prior art in this area includes US 4,835,617,  
3 US 5,576,762, US 5,134,489, US 5,122,881, US  
4 5,471,515 and WO 98/08079.

5

6 An object of the present invention is to provide a  
7 solid state image sensor which, like the 3T sensor,  
8 can be manufactured by standard techniques, but which  
9 also is capable of providing a true electronic  
10 shutter.

11

12 The invention and preferred features thereof are  
13 defined in the appended Claims.

14

15 Briefly stated, the invention is based upon locating  
16 the readout electronics off the image plane of the  
17 device. In preferred forms of the invention, this is  
18 facilitated by connecting each pixel to its  
19 associated readout electronics via a multi-conductor  
20 signal bus.

21

22 Embodiments of the invention will now be described,  
23 by way of example only, referring to the drawings in  
24 which:

25

26 Figures 1 to 5 illustrate the prior art discussed  
27 above;

28

29 Figure 6 shows a part of one column of an array  
30 structure embodying the invention;

31

1 Figure 7 is a timing diagram illustrating the  
2 operation of Figure 6;

3  
4 Figure 8 shows a typical system layout of a sensor  
5 incorporating the circuitry of Figure 6;

6  
7 Fig. 9 shows one pixel and read-out circuitry of a  
8 modified version of Fig. 6;

9  
10 Figure 10 is a timing diagram illustrating the  
11 operation of Figure 9;

12  
13 Figure 11 shows one pixel plus read-out circuitry of  
14 a further modification of Fig. 6;

15  
16 Figure 12 is a timing diagram illustrating the  
17 operation of Figure 11;

18  
19 Figure 13 is a view similar to Figure 8 but showing a  
20 modified system layout; and

21  
22 Figure 14 shows a preferred readout arrangement for  
23 the circuit of Figure 11.

24  
25 A basic feature of the invention is to provide a  
26 storage node per pixel and, to avoid degrading the  
27 fill factor (and hence light sensitivity), locating  
28 the storage element away from the image plane.

29  
30 Referring to Figure 6, this embodiment has only two  
31 transistors, M1 and M2, per pixel, thus improving the

1 fill factor and sensitivity. The array is not  
2 multiplexed and therefore there is no multiplex  
3 transistor in the pixel equivalent to M3 in Figure 1.  
4 Instead, there is a connection to the signal bus 10  
5 which runs through the column.

6  
7 The switches S2-1, S2-2 etc will typically be  
8 implemented as MOSFET transistors. The current loads  
9 Iload are to ensure correct operation of sense  
10 transistor M1. Figure 6 shows only two pixels, but  
11 in a practical array there are several pixels in a  
12 column.

13  
14 The operation of the array is as follows. At point 1  
15 (see Figure 7) the RST signal goes high, causing all  
16 the "M2" transistors (M2\_1, M2\_2 etc) to conduct and  
17 the voltage Vpix on the photodiode to be reset to  
18 Vrt. At a time later, point 2 (see Figure 7), all  
19 the "S1" switches (S1\_1, S1\_2 etc) are closed  
20 simultaneously and the output of the sense  
21 transistors (M1) are stored on the sense capacitors  
22 (Csn\_1, Csn\_2). Subsequently (not shown), the  
23 signals on the sense capacitors are readout  
24 sequentially by sequentially closing switches S2  
25 (S2\_1, S2\_2 etc).

26  
27 Figure 8 shows a typical layout of a system, with an  
28 image array 12 and sample capacitor area 14. For  
29 ease of drawing, a 6 x 6 pixel structure is shown but  
30 the array would typically be larger. Note how the  
31 output from each pixel is wired ("X" in Figure 8) to

1 a different conductor of the signal bus 10. Note  
2 also the cell width A of the system.

3  
4 The embodiment of Figures 6 to 8 shows signal bus  
5 lines planar with the image plane, i.e. using the  
6 same conductor layer. One improvement (not shown) is  
7 to stack the conductors, that is to use different  
8 conductive layers. This reduces the amount of metal  
9 covering the pixel and thus improves the amount of  
10 light collected by the pixel.

11  
12 The system described in Figure 6 is area and cost  
13 efficient but it suffers from "Fixed Pattern Noise"  
14 in the form of brightness variations on the picture.  
15 This is due to the varying amount of "threshold  
16 voltage" of transistors M1 over the array. These  
17 variations are a normal part of CMOS manufacturing  
18 process. A practical way of cancelling this offset  
19 is to measure, on a per-pixel basis, the reset  
20 voltage after the source follower.

21  
22 Referring to Figures 9 and 10, this is achieved by  
23 closing switch S3 (Figure 9) immediately after the  
24 end of the reset pulse ("2" in Figure 10). This  
25 signal is then stored on "Cres" and switch S3 is  
26 opened. For a period of time ("3" in Figure 10), the  
27 pixel collects light and the photo-charge discharges  
28 the photodiode. At the end of this period ("4" in  
29 Figure 10) the signal is sampled on "Csn". During  
30 image readout ("5" in Figure 10), switches S2 and S4  
31 are closed simultaneously and both the signal and

1 reset values are output onto the "Output Signal" and  
2 "Reset Value" conductors. The threshold voltage can  
3 then be compensated by subtracting the "Reset Value"  
4 from the "Output Signal".

5

6 This technique is similar to that used in US  
7 5,122,881 but is modified to deal with the present  
8 situation where no multiplex transistor is present.

9

10 Although the technique described previously (Figure  
11 9) cancels the offset, it degrades the rate at which  
12 the system can operate as it is not possible to  
13 perform image acquisition and readout simultaneously.  
14 This is because the reset signal ("2" in Figure 10)  
15 occurs at the start of an image acquisition, but is  
16 required during readout. A new acquisition is  
17 therefore not possible until readout has been  
18 completed.

19

20 The solution to this problem is shown in Figure 11.  
21 An extra capacitor per pixel is used to enable  
22 simultaneous image acquisition and readout.

23

24 To understand the operation of the circuit in Figure  
25 11, refer to the timing diagram in Figure 12:

26

- 27 • At point "1", Vrst goes high causing all the  
28 M2s in the array to conduct, resetting the  
29 photodiodes in the array.

- 1       • As soon as this is complete, (point "2") S2  
2       goes high enabling CresA to sample the reset  
3       value of the pixel.
- 4       • The image array collects light until time "3"  
5       when the voltage corresponding to the pixel's  
6       exposure to light is collected. S1 is closed  
7       and the voltage is stored on the pixel's Csn.

8  
9       At this time the system has collected a complete set  
10      of reset and image values and is ready to readout.  
11      Before this occurs, the next acquisition cycle  
12      starts:

- 13       • At point "4", Vrst goes high causing all the  
14       M2s in the array to conduct, resetting the  
15       photodiodes in the array.
- 16       • As soon as this is complete, (point "5") S4  
17       goes high enabling CresB to sample the reset  
18       value of the pixel.
- 19       • As the image array collects light, the  
20       pixels' capacitors are accessed sequentially.  
21       At point "6", S2 is closed to output the  
22       image value "Vsn" stored on Csn onto the  
23       "Output Signal" conductor. For this sequence  
24       of images, S4 is closed to output the reset  
25       value "Vres" stored on CresA onto the "Reset  
26       Value A" conductor.
- 27       • The image array collects light until time "7"  
28       when the voltage corresponding to the pixel's  
29       exposure to light is collected. S1 is closed  
30       and the voltage is stored on the pixel's Csn.

1

2 At this time the system has collected another  
3 complete set of reset and image values and is ready  
4 to readout. Before this occurs, the next acquisition  
5 cycle starts:

- 6 • Point "8" is identical to point "1"
- 7 • Point "9" is identical to point "2"
- 8 • As the image array collects light, the  
9 pixels' capacitors are accessed sequentially.  
10 At point "10", S2 is closed to output the  
11 image value "Vsn" stored on Csn onto the  
12 "Output Signal" conductor. For this sequence  
13 of images, S6 is closed to output the reset  
14 value "Vres" stored on CresB onto the "Reset  
15 Value B" conductor.

16

17 The system continues to operate using the sequence  
18 described above. The important feature to note on  
19 Figure 12 is that Vsn is able to be output on each  
20 frame.

21

22 In the layout shown in Figure 8, the pitch of the  
23 sample capacitors is  $1/6^{\text{th}}$  the pitch of the pixels as  
24 there are 6 pixels vertically. For a larger array, a  
25 greater number of sample capacitors needs to be  
26 fitted into the width of a pixel. This presents a  
27 practical limit to the architecture - the minimum  
28 width of sample capacitors is determined by the  
29 manufacturing technology used by the architecture,  
30 the maximum size of the pixel is determined by cost  
31 factors.

An improved layout is shown in Figure 13. This architecture has sample capacitors 14A and 14B at the top and bottom of the array 12. There are now two signal buses 10A and 10B, divided in the centre, and the cell width B is equal to  $1/3$  of a pixel. There are two advantages.

- (1) The fewer signal bus conductors running across each pixel requires less metal and hence there is less obstruction of the pixel (i.e higher fill-factor) and hence greater sensitivity from the pixel.
- (2) As the array is divided into two parts, the sample capacitors are shared top and bottom, resulting in twice the width available.

The following table illustrates the advantages:

Layout	Column Width	Pixel Array	Pixel Size	Image Plane	Imaging Area
Fig. 8	$2\mu\text{m}$	100x100	$200\mu\text{m} \times 200\mu\text{m}$	$200\text{mm} \times 200\text{mm}$	$400\text{m}^2$
Fig. 13	$2\mu\text{m}$	100x100	$100\mu\text{m} \times 100\mu\text{m}$	$100\text{mm} \times 10\text{mm}$	$100\text{m}^2$

As can be seen in the final column, the improved layout technique of Fig. 13 produces a four-fold increase in area (and hence a corresponding reduction in cost per unit area).



1 Turning to Figure 14, a preferred scheme for  
2 measuring and amplifying the two output signals will  
3 now be described.

4  
5 Associated with the switches S2, S4, S6 and the  
6 conductors "Output Signal" 18, "Reset Value A" 20,  
7 and "Reset Value B" 22, are unwanted stray  
8 capacitances. As the array size increases, the  
9 number of pixels and therefore the number of switches  
10 increases. The cumulation of all these switches can  
11 produce an unwanted capacitance roughly equal to that  
12 of the sampling capacitances. When the signals are  
13 read out (switches S2/S4/S6 closed), part of the  
14 charge stored on the capacitors Csn/CresA/CresB is  
15 used to charge the stray capacitors. This problem is  
16 known as "charge sharing". This can easily be 50% to  
17 70% of the signal, reducing the output swing to 1/2  
18 or 1/4 of the "true" signal.

19  
20 Using a differential, charge sensitive amplifier 16  
21 as shown in Figure 14 charge sharing is avoided.  
22 Before the signal is read out, the switches S7, S8  
23 are closed and the amplifier 16 put into its "common  
24 mode reset" state. This discharges the capacitors  
25 Cf1, Cf2 on the feedback of the operational amplifier  
26 16 and forces the conductors 18, 20, 22 to the common  
27 mode voltage. Switches S7/S8 are opened and S2, S4  
28 (or S6) are then closed. The nature of the  
29 operational amplifier is to ensure that its input  
30 remains at the common mode voltage. By doing so  
31 there is no change in voltage on the lines 18, 20 and

1 22 and so there can be no loss of charge. During the  
2 readout, the voltages on Csn, CresA, CresB are also  
3 set to the common mode voltage. The change in  
4 voltage from that which was measured off the array  
5 requires a current to flow. This comes from the  
6 output of the op-amp 16 via the feedback capacitors  
7 Cf1, Cf2. For correct (symmetrical operation) the  
8 capacitance of Cf1 = Cf2 and Csn=CresA=CresB. Hence:

9

$$10 \quad \text{Out1} - \text{Out2} = (\text{Vsignal} - \text{Vreset}) \times \text{Csn} / \text{Cf1}$$

11

12 Modifications and improvements may be made to the  
13 foregoing within the scope of the invention.

14

15

1     Claims

2

3     1.    A solid state imaging device comprising a two-  
4           dimensional array of pixels forming an image  
5           plane, and readout electronics for reading out  
6           signals from the pixels in a predetermined  
7           manner; and in which the readout electronics are  
8           located off said image plane.

9

10    2.    The device of claim 1, in which each pixel is  
11           connected to its associated readout electronics  
12           via a multiconductor signal bus.

13

14    3.    The device of Claim 2, in which each pixel  
15           comprises a photosensitive diode and switching  
16           means for resetting and discharging the diode;  
17           and in which the switching means consists only  
18           of a first transistor for applying a reset pulse  
19           and a second transistor operable to connect the  
20           diode to a predetermined conductor of said  
21           multi-conductor signal bus.

22

23    4.    The device of Claim 2 or Claim 3, in which the  
24           signal bus conductors are stacked.

25

26    5.    The device of any preceding Claim, in which the  
27           readout electronics are located at one side of  
28           the array.

29

- 1     6.    The device of any one of Claims 1 to 5, in which  
2           the readout electronics are located on two  
3           opposite sides of the array.  
4
- 5     7.    The device of any preceding Claim, in which all  
6           pixels in the array are reset simultaneously and  
7           are read out simultaneously.  
8
- 9     8.    The device of any preceding Claim, in which the  
10          readout electronics comprises, for each pixel, a  
11          first store for a reset value and a second store  
12          for a read out value; and the readout  
13          electronics is effective to modify the read out  
14          value of a given pixel by the stored reset value  
15          for that pixel.  
16
- 17    9.    The device of Claim 8, in which the readout  
18          electronics further includes, for each pixel, a  
19          further store for a second reset value whereby  
20          the current reset and read out values may be  
21          processed simultaneously with applying a new  
22          reset pulse.  
23
- 24    10.   The device of Claim 9, in which the readout  
25          electronics further includes a differential  
26          amplifier connectable to said stores, and means  
27          for putting the amplifier into a common mode  
28          reset state prior to reading out a signal.  
29

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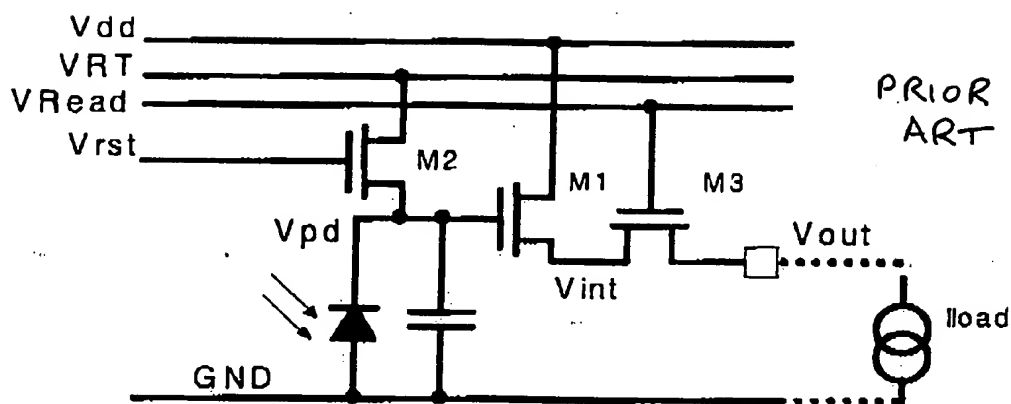


Figure 1 Three Transistor Pixel

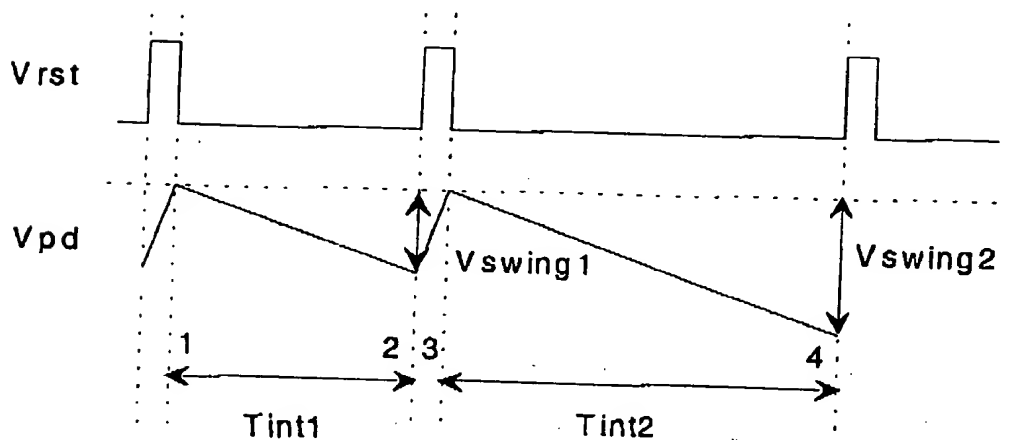


Figure 2 Voltage Swing on Photodiode

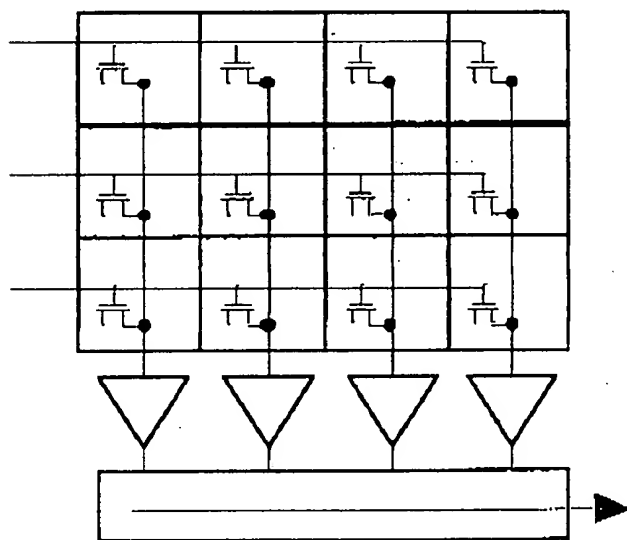
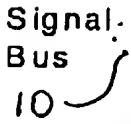


Figure 3 Multiplex Readout



### Figure 5 Sophisticated Linear Array



### Output Signal



### Figure 7 Timing Diagram

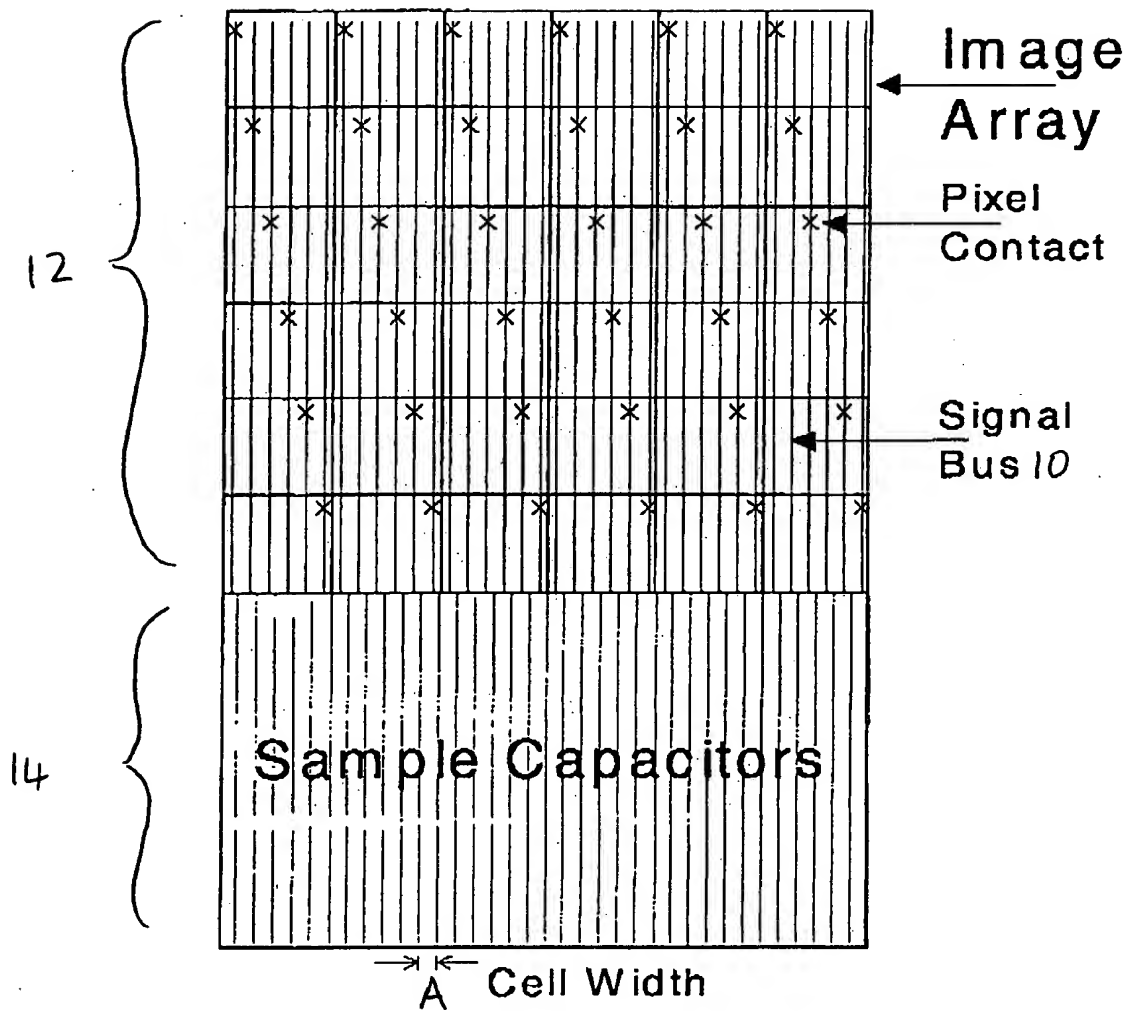


Figure 8 Typical System Layout



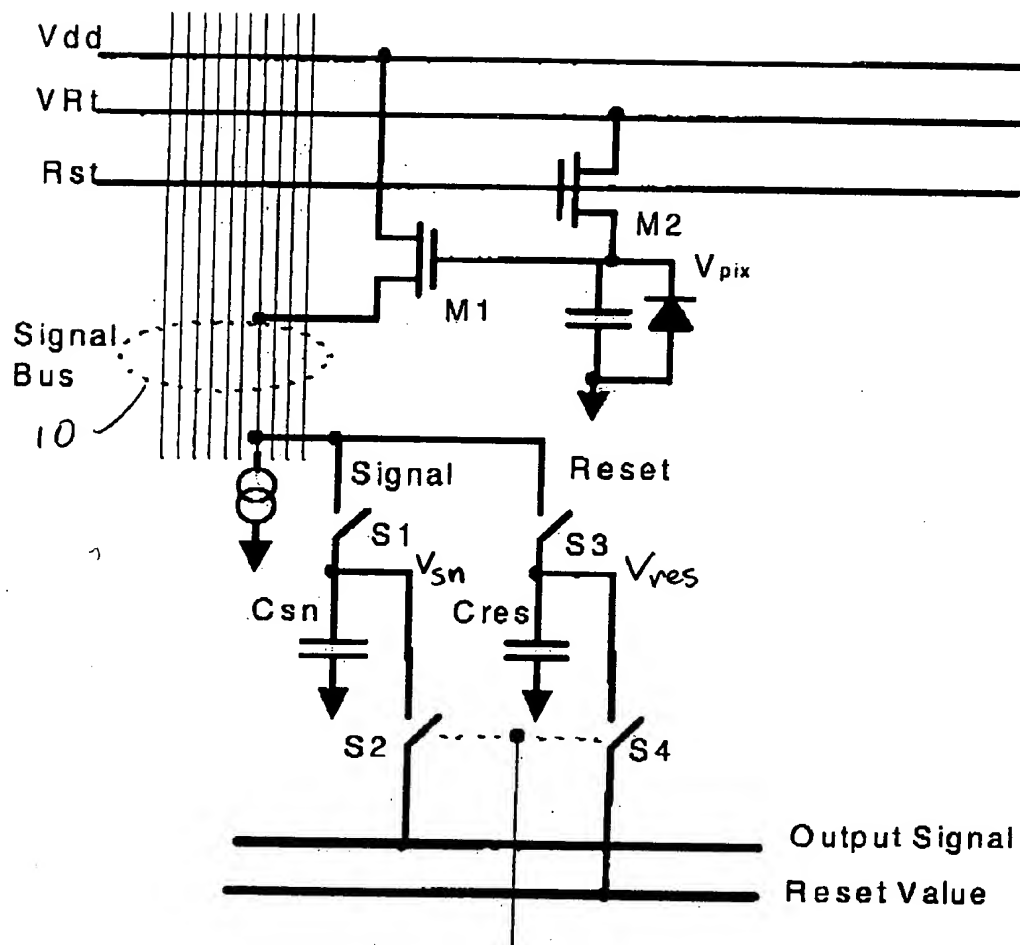


Figure 9 Improved Circuit - Offset Cancellation

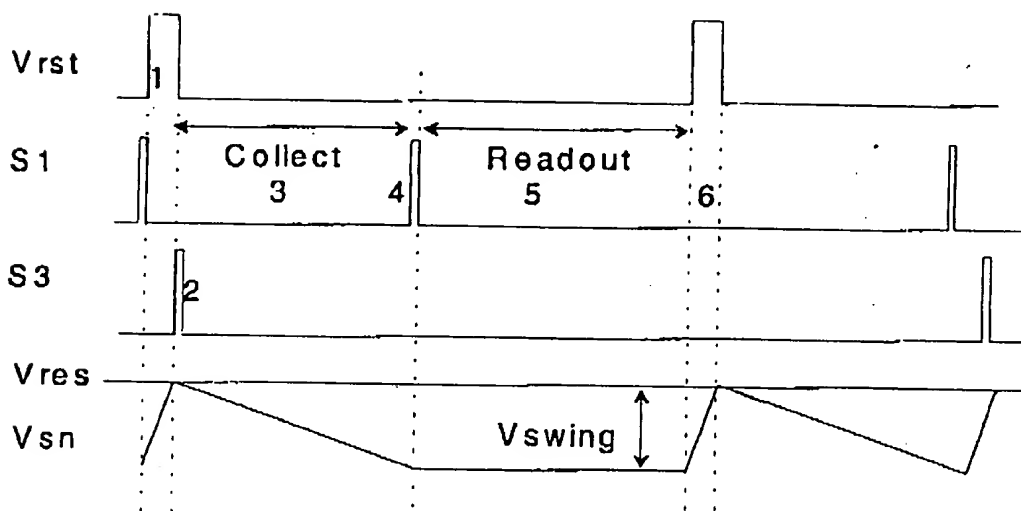


Figure 10 Offset Cancellation - Timing diagram

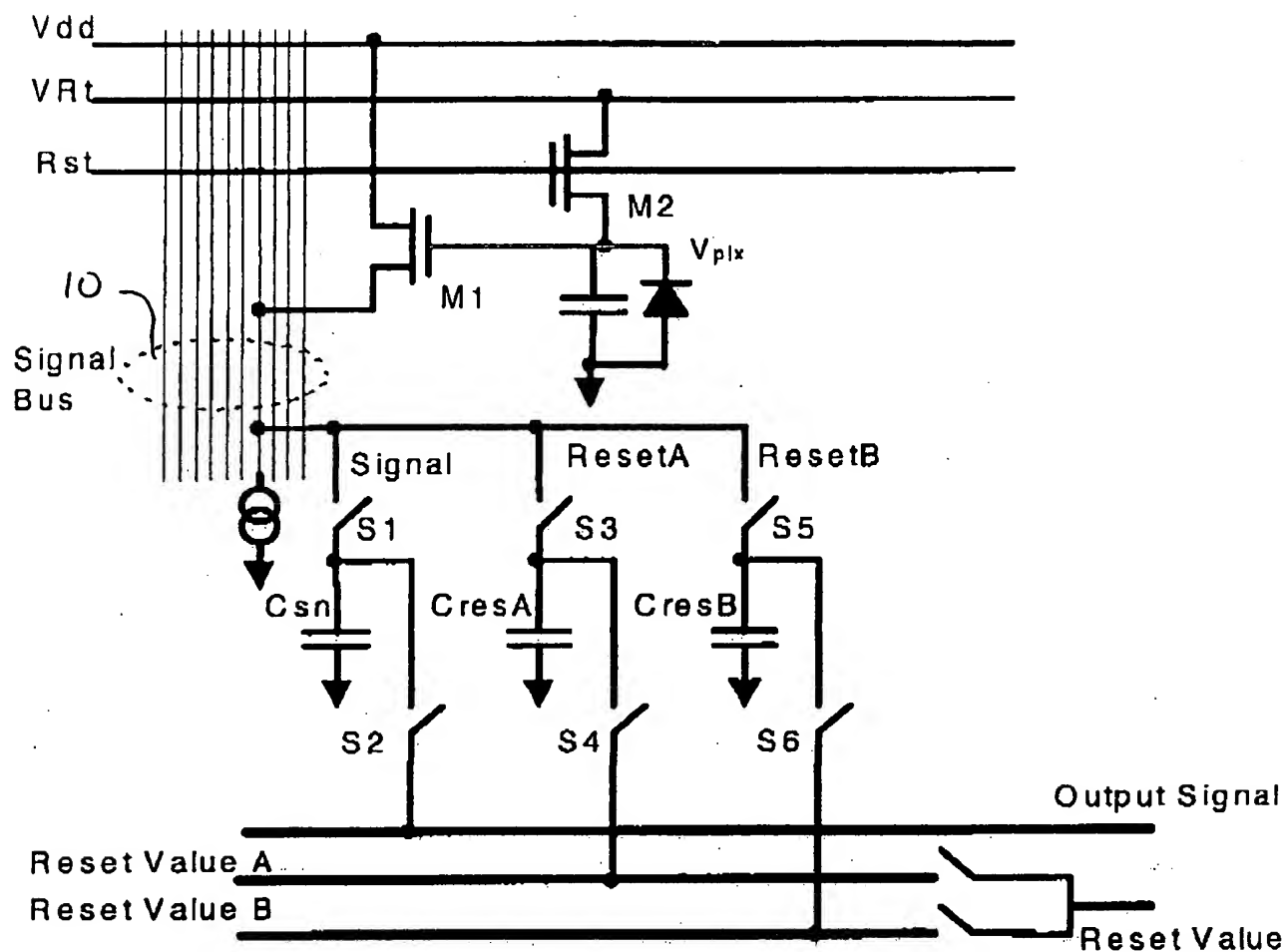


Figure 11 Improved Circuit - Offset Cancellation 2

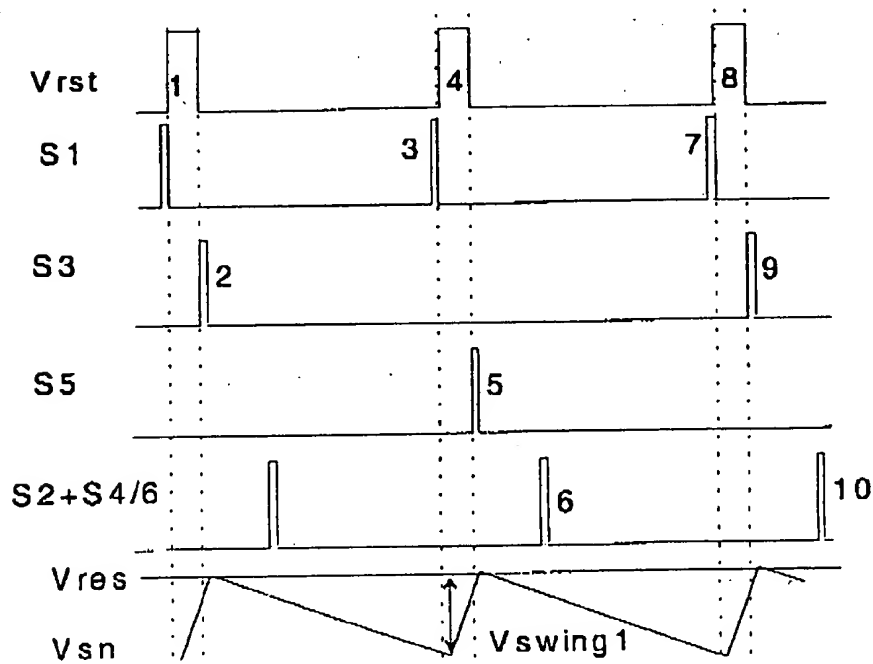


Figure 12 Offset Compensation 2 - Timing Diagram

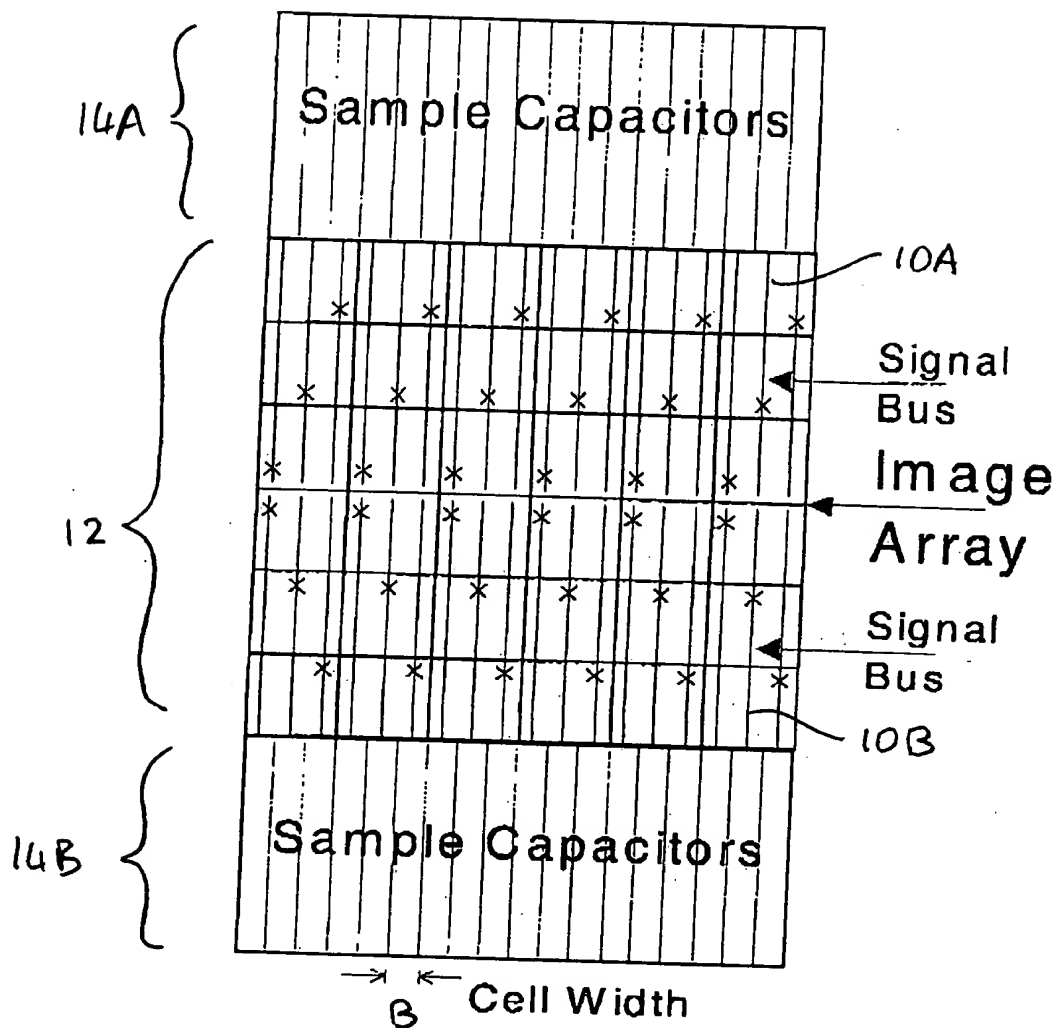


Figure 13 Improved Layout Technique

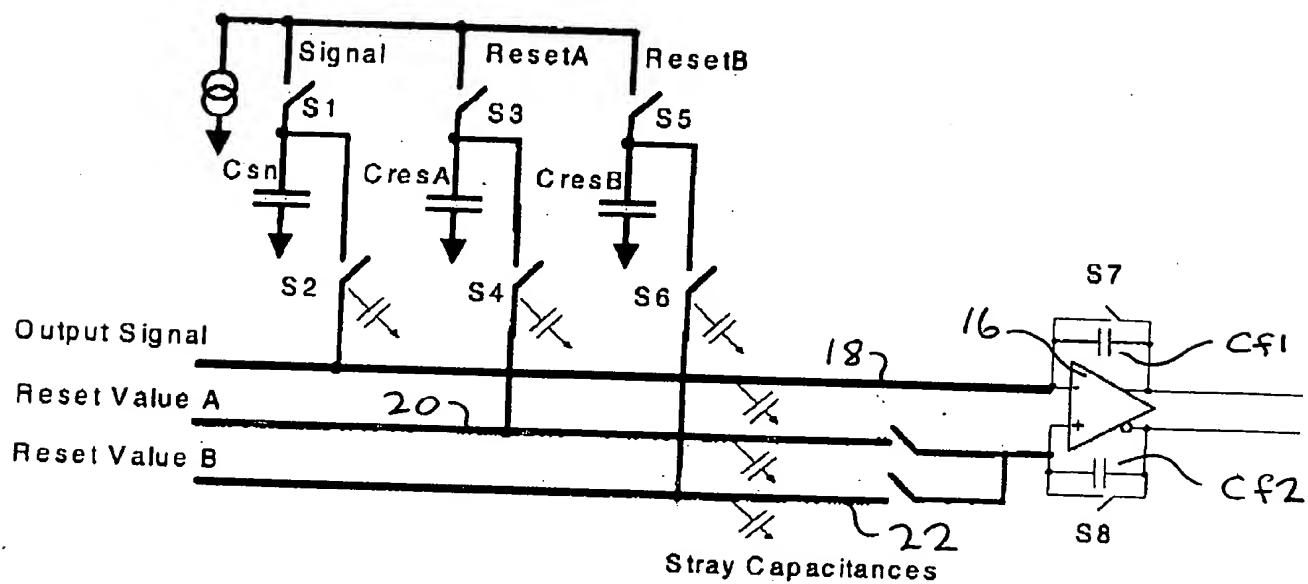


Figure 14 Preferred Readout Amplification